### Control Systems I

Linear Quadratic Regulator

Colin Iones

Laboratoire d'Automatique

### Motivation for LQR

Consider the system

$$\dot{x} = Ax + Bu$$

$$y = Cx$$

Define  $Q = C^T C$  and  $R = \rho I$ . We are minimizing the cost

$$V(x, u) = \int_0^\infty x(t)^T Qx(t) + u(t)^T Ru(t) dt$$
$$= \int_0^\infty y(t)^2 + \rho u(t)^2$$

We're minimizing the relative  $\emph{energy}$  in the input and output signals

Large  $ho 
ightarrow {
m small}$  input energy, output weakly controlled Small  $ho 
ightarrow {
m large}$  input energy, output strongly controlled

Note: Any minimal solution must be stable / take the state to the origin. Why? Any non-zero steady-state solution will result in an infinite cost V(x,u)

## Linear Quadratic Regulator

**Goal:** Move from state x to the origin. (i.e., keep x 'small')

$$\dot{x} = Ax + Bu$$

Express the 'cost' of being in state x and applying input u with the function

$$l(x, u) = x^T Q x + u^T R u$$

The total 'cost' of following a particular trajectory is then

$$V(x, u) = \int_0^\infty x(t)^T Qx(t) + u(t)^T Ru(t) dt$$

Assume:  $R \succ 0$ ,  $Q \succeq 0$  (i.e., R is positive definite, and Q is positive semi-definite)

### LQR: Find the 'best' trajectory

$$\min_{u} V(x(t), u(t))$$
 s.t. 
$$\dot{x}(t) = Ax(t) + Bu(t)$$

2

### Motivation for LQR

Consider the system

$$\dot{x} = Ax + Bu$$

$$y = Cx$$

Define  $Q = C^T C$  and  $R = \rho I$ . We are minimizing the cost

$$V(x,u) = \int_0^\infty x(t)^T Qx(t) + u(t)^T Ru(t) dt$$
$$= \int_0^\infty y(t)^2 + \rho u(t)^2$$

We're minimizing the relative *energy* in the input and output signals

Real motivation

- · Works well in practice
- Works seamlessly for multi-input, multi-output systems
- We can solve it (very common motivation in control!)
- · Solution is simple, and easy to implement in embedded controller

3

3

#### LQR - Solution

#### Linear Quadratic Regulator

Consider a linear multivariable system  $\dot{x}(t)=Ax(t)+Bu(t)$ . Compute control law u(t)=-Kx(t) such that the following performance criterion is minimized

$$\min_{u} \int_{0}^{\infty} x(t)^{T} Q x(t) + u(t)^{T} R u(t) dt$$
s.t.  $\dot{x}(t) = A x(t) + B u(t)$ 

where R is positive definite, and Q is positive semi-definite.

#### **Optimal Solution**

The optimal controller is u(t) = -Kx(t) where

$$K = -R^{-1}B^T P$$

and  $P = P^T \succ 0$  is the solution of the following **Riccati Equation** 

$$A^T P + PA - PBR^{-1}B^T P + Q = 0$$

### Feedback Invariants

A quantity is called a *feedback invariant* if its value does not depend on the choice of the control input  $u(t), t \ge 0$ .

#### Lemma: Feedback Invariant

Let P be a symmetric matrix. For every control input u(t),  $t\in [0,\infty]$  for which  $x(t)\to 0$  as  $t\to \infty$ , we have that

$$\int_{0}^{\infty} x^{T} (A^{T} P + PA) x + 2x^{T} PB u dt = -x(0)^{T} Px(0)$$

Proof:

$$\int_0^\infty x^T (A^T P + PA)x + 2x^T PBu dt = \int_0^\infty (x^T A^T + u^T B^T) Px + x^T P (Ax + Bu) dt$$

$$= \int_0^\infty \dot{x}^T Px + x^T P \dot{x} dt$$

$$= \int_0^\infty \frac{d(x^T Px)}{dt} dt$$

$$= \lim_{t \to \infty} x^T (t) Px (t) - x(0)^T Px (0)$$

#### **Proof Sketch**

We want to minimize the function

$$J_{LQR} = \int_0^\infty x(t)^T Qx(t) + u(t)^T Ru(t) dt$$

Idea<sup>1</sup>:

We will first write it as

$$J_{LQR} = J_0 + \int_0^\infty (u(t) - u_0(t))^T R(u(t) - u_0(t)) dt$$

for some  $u_0$ , where  $J_0$  does not depend on the control law chosen.

From this we can see that the optimal input is  $u(t) = u_0(t)$ .

5

# **Square Completion**

$$J_{LQR} = \int_0^\infty x(t)^T Qx(t) + u(t)^T Ru(t) dt$$

Add and subtract our feedback invariant

$$= x(0)^{T} P x(0) + \int_{0}^{\infty} x^{T} Q x + u^{T} R u + x^{T} (A^{T} P + P A) x + 2x^{T} P B u dt$$

Re-write the terms involving u

$$u^{T}Ru + 2x^{T}PBu = (u - u_{0})^{T}R(u - u_{0}) - x^{T}PBR^{-1}B^{T}Px$$

where  $u_0 = -R^{-1}B^TPx$ 

Which gives us

$$J_{LQR} = x(0)^{T} P x(0) + \int_{0}^{\infty} x^{T} Q x + x^{T} (A^{T} P + P A) x - x^{T} P B R^{-1} B^{T} P x dt$$
$$+ \int_{0}^{\infty} (u - u_{0})^{T} R (u - u_{0}) dt$$

<sup>&</sup>lt;sup>1</sup>We're following the proof of Joao P. Hespanha to avoid variational analysis

## **Square Completion**

$$J_{LQR} = x(0)^{T} P x(0) + \int_{0}^{\infty} x^{T} Q x + x^{T} (A^{T} P + P A) x - x^{T} P B R^{-1} B^{T} P x dt$$
$$+ \int_{0}^{\infty} (u - u_{0})^{T} R (u - u_{0}) dt$$

We see that the optimal solution is

$$u = u_0 = -Kx$$

$$K = R^{-1}B^T P$$

and

$$Q + A^{T}P + PA - PBR^{-1}B^{T}P = 0$$

8

## Example

Solve the ARE  $A^TP + PA - PBR^{-1}B^TP + Q = 0$ 

$$\begin{bmatrix} 0 & 0 \\ 1 & 0 \end{bmatrix}^{T} \begin{bmatrix} P_{1} & P_{2} \\ P_{2} & P_{3} \end{bmatrix} + \begin{bmatrix} P_{1} & P_{2} \\ P_{2} & P_{3} \end{bmatrix} \begin{bmatrix} 0 & 0 \\ 1 & 0 \end{bmatrix} - \begin{bmatrix} P_{1} & P_{2} \\ P_{2} & P_{3} \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix}^{T} \begin{bmatrix} P_{1} & P_{2} \\ P_{2} & P_{3} \end{bmatrix} + \begin{bmatrix} 0 & 0 \\ 0 & q^{4} \end{bmatrix} = 0$$

This results in four equations

$$-P_1^2 + 2P_2 = 0$$

$$P_3 - P_1P_2 = 0$$

$$P_3 - P_1P_2 = 0$$

$$q^4 - P_2^2 = 0$$

### Example

Compute linear controller to minimize the closed-loop performance metric

$$Q = q^4 C^T C$$

$$R = 1$$

for the system  $G(s) = \frac{1}{s^2}$ , whose state-space representation is

$$\dot{x} = \begin{bmatrix} 0 & 0 \\ 1 & 0 \end{bmatrix} x + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u$$

$$y = \begin{bmatrix} 0 & 1 \end{bmatrix} x$$

.

## Example

Solving gives

$$P = \begin{bmatrix} \sqrt{2}q & q^2 \\ q^2 & \sqrt{2}q^3 \end{bmatrix}$$

(Note that we've selected the real, positive definite solution)

The controller is

$$K = R^{-1}B^TP = \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} \sqrt{2}q & q^2 \\ q^2 & \sqrt{2}q^3 \end{bmatrix} = \begin{bmatrix} \sqrt{2}q & q^2 \end{bmatrix}$$

## Closed-loop System

The closed-loop system is

$$\dot{x} = (A - BK)x = \begin{bmatrix} -\sqrt{2}q & -q^2 \\ 1 & 0 \end{bmatrix} x$$
$$y = \begin{bmatrix} 0 & 1 \end{bmatrix} x$$

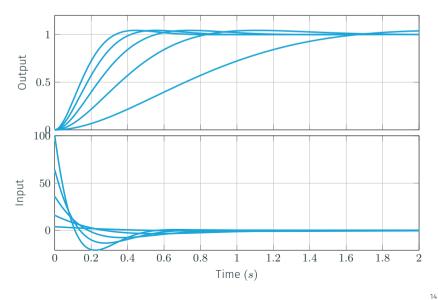
Which has poles at  $q\frac{1}{\sqrt{2}}(-1\pm i)$ 

We see that

- As  $q \to 0$ , the input energy is dominant, and the closed-loop poles become the open-loop poles
- As *q* becomes large, the output energy is dominant, and the system spends more input energy to bring the input to zero faster

12

# Step Response



## Add Reference Input

Add a reference input to the system

$$\dot{x} = \begin{bmatrix} -\sqrt{2}q & -q^2 \\ 1 & 0 \end{bmatrix} x + \begin{bmatrix} 1 \\ 0 \end{bmatrix} \bar{N}r$$

$$y = \begin{bmatrix} 0 & 1 \end{bmatrix} x$$

We want to steady-state gain between r and y to be one (i.e., y = r at steady-state)

$$0 = \begin{bmatrix} -\sqrt{2}q & -q^2 \\ 1 & 0 \end{bmatrix} x + \begin{bmatrix} 1 \\ 0 \end{bmatrix} \bar{N}r \to x = -\begin{bmatrix} -\sqrt{2}q & -q^2 \\ 1 & 0 \end{bmatrix}^{-1} \begin{bmatrix} 1 \\ 0 \end{bmatrix} \bar{N}r$$
$$y = \begin{bmatrix} 0 & 1 \end{bmatrix} x = -\begin{bmatrix} 0 & 1 \end{bmatrix} \begin{bmatrix} -\sqrt{2}q & -q^2 \\ 1 & 0 \end{bmatrix}^{-1} \begin{bmatrix} 1 \\ 0 \end{bmatrix} \bar{N}r$$

We choose  $\bar{N}$  such that y=r

$$1 = -\begin{bmatrix} 0 & 1 \end{bmatrix} \begin{bmatrix} -\sqrt{2}q & -q^2 \\ 1 & 0 \end{bmatrix}^{-1} \begin{bmatrix} 1 \\ 0 \end{bmatrix} \bar{N}$$
$$= \frac{1}{q^2} \bar{N}$$

So  $\bar{N}=q^2$ 

# **Choice of Weights**

Weights are usually determined through a trial-and-error process, but a good initial setting is given by Bryson's rule.

Bryson's rule scales the variables so that the maximum acceptable value for each term is one.

Choose diagonal Q and R with

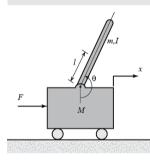
$$Q_{ii} = rac{1}{ ext{maximum acceptable value of } x_i^2} \ R_{jj} = rac{1}{ ext{maximum acceptable value of } u_j^2}$$

Start with Bryson's rule, and then increase or decrease the diagonal values to increase or decrease the convergence rates of the corresponding states.

13

## LQR in Matlab

$$\begin{bmatrix} \dot{x} \\ \ddot{x} \\ \dot{\theta} \\ \ddot{\theta} \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & -0.1818 & 2.6727 & 0 \\ 0 & 0 & 0 & 1 \\ 0 & -0.4545 & 31.1818 & 0 \end{bmatrix} \begin{bmatrix} x \\ \dot{x} \\ \theta \\ \dot{\theta} \end{bmatrix} + \begin{bmatrix} 0 \\ 1.8182 \\ 0 \\ 4.5455 \end{bmatrix}$$
$$y = \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 \end{bmatrix} \begin{bmatrix} x \\ \dot{x} \\ \theta \\ \dot{\theta} \end{bmatrix}$$



Objectives for a 0.2m step in cart position x are:

- Settling time for x and  $\theta$  of less than 5 seconds
- Rise time for x of less than 0.5 seconds
- Pendulum angle  $\theta$  never more than 20 degrees (0.35 radians) from the vertical
- Steady-state error of less than 2% for x and  $\theta$

## Conclusion

- · Define the behaviour that we want to achieve via a value function
- Choose the control law that minimizes the value function
- LQR is very effective because:
  - The optimal controller is linear
  - The value function is intuitive to tune